

RF Power Amplifier Design

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Contents

- ⊙ Basic Amplifier Concepts
 - Class A, B, C, F, hHCA
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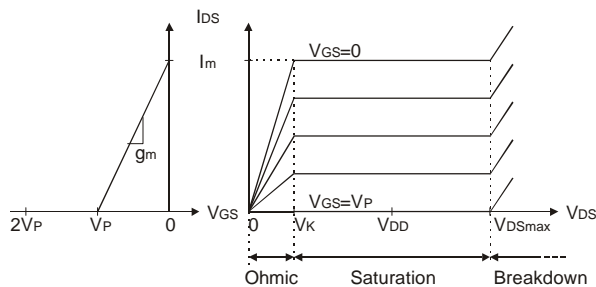
- ⊙ Enhanced Amplifier Concepts
 - Feedback, Feedforward, ...
 - Predistortion
 - LINC, Doherty, EER, ...

Efficiency Definitions

⊙ Drain Efficiency: $h_D = \frac{P_{OUT}}{P_{DC}}$

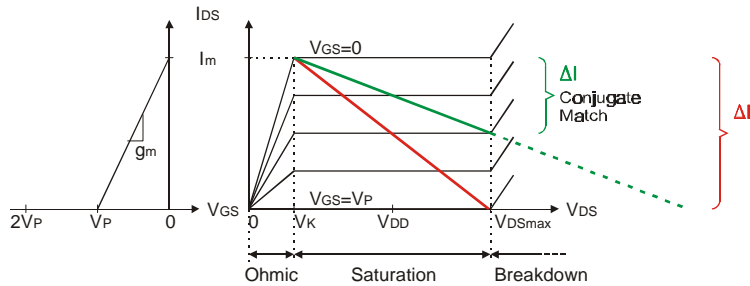
⊙ Power Added Efficiency: $h_{PA} = \frac{P_{OUT} - P_{IN}}{P_{DC}} = h_D \cdot \left(1 - \frac{1}{G}\right)$

Ideal FET Input and Output Characteristics



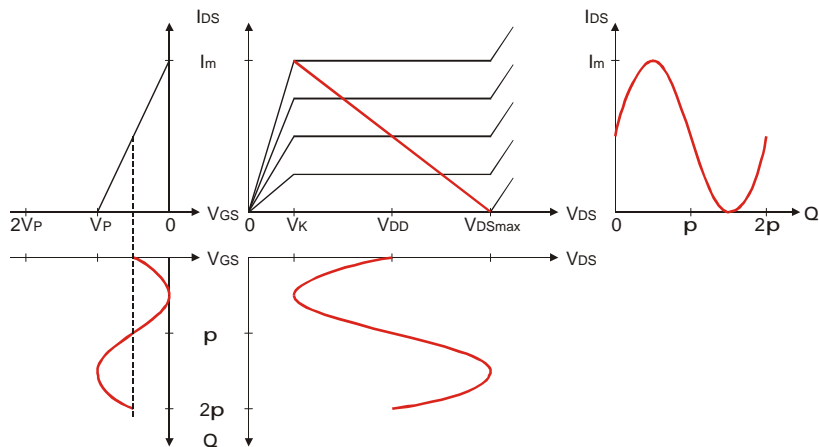
$$k = \frac{V_{DD} - V_K}{V_{DD}}$$

Maximum Output Power Match

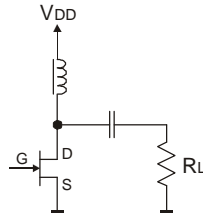


$$R_{OPT} = \frac{V_{DSmax} - V_K}{I_m}$$

Class A



Class A - Circuit

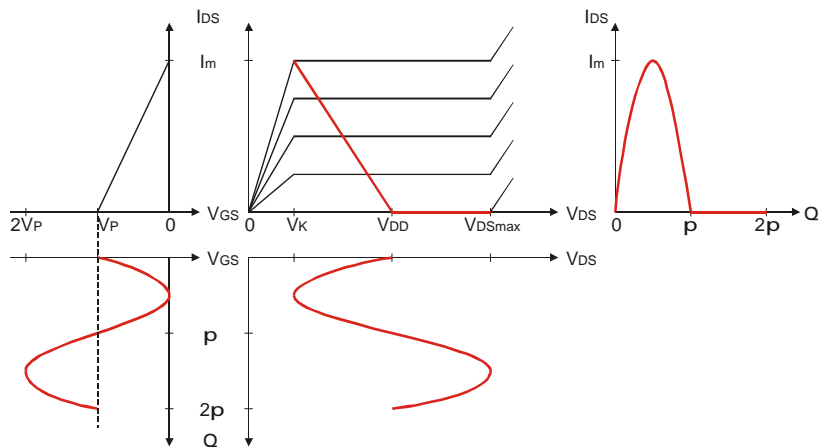


$$h_D = k \cdot 50\%$$

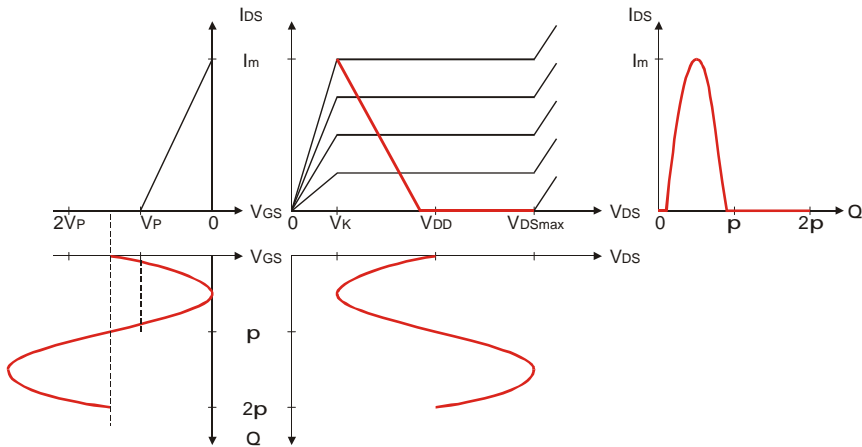
$$G = G_A \text{ (e.g. 14 dB)}$$

$$h_{PA} = k \cdot 48\%$$

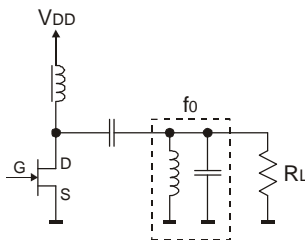
Class B



Class C



Class B and C – Circuit



Class B

$$h_d = k \cdot 78\%$$

$$G = G_A - 6\text{dB} \quad (8\text{ dB})$$

$$h_{pA} = k \cdot 65\%$$

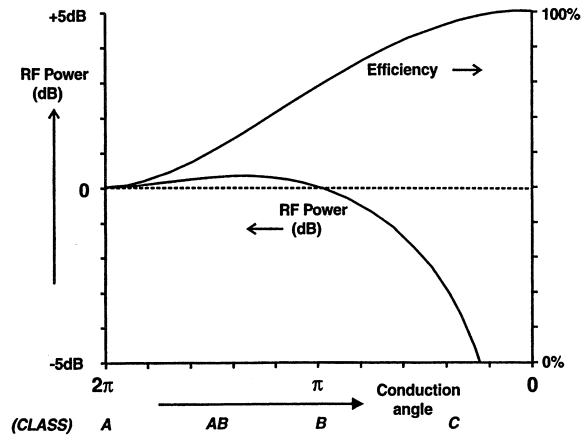
Class C

$$h_d \rightarrow 100\%$$

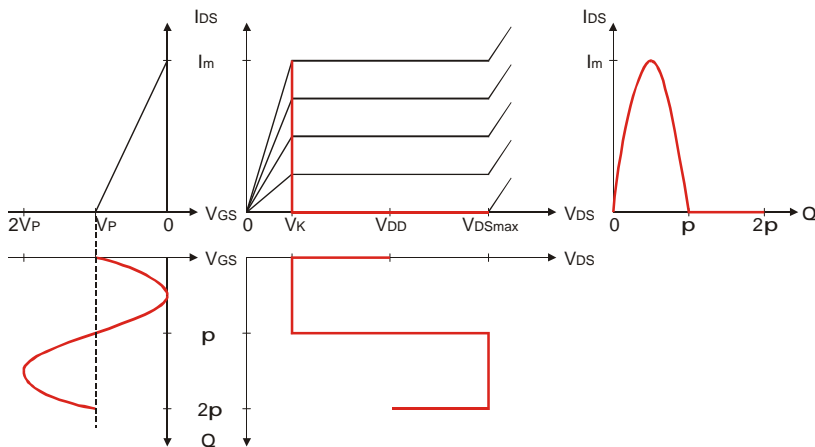
$$G \rightarrow 1$$

$$h_{pA} \rightarrow 0\%$$

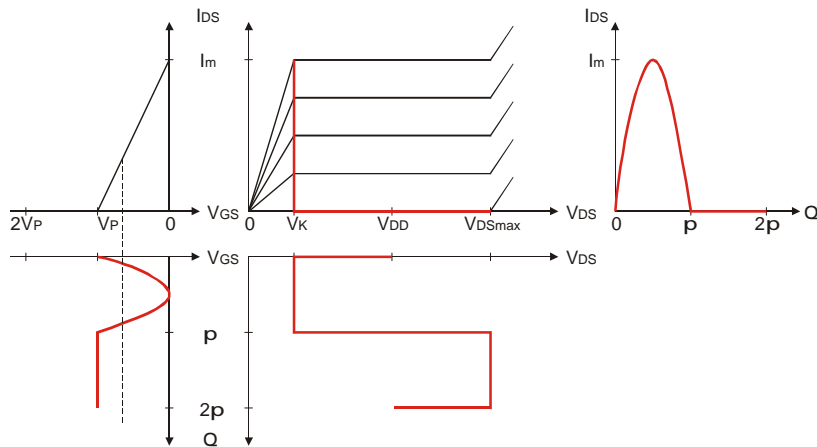
Influence of Conduction Angle



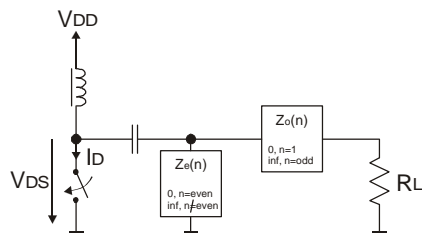
Class F (HCA ... harmonic controlled amplifier)



hHCA (half sinusoidally driven HCA)



Class F and hHCA - Circuit



Class F

$$h_D = k \cdot 100\%$$

$$G = G_A - 5\text{dB} \quad (9\text{ dB})$$

$$h_{PA} = k \cdot 87\%$$

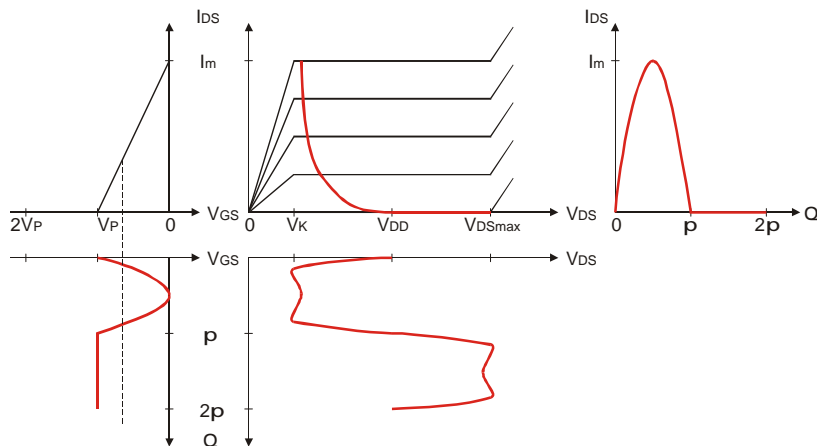
hHCA

$$h_D = k \cdot 100\%$$

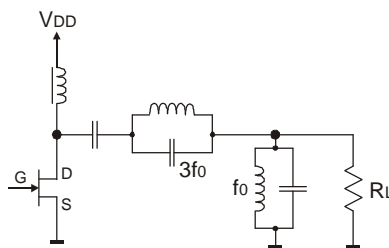
$$G = G_A + 1\text{dB} \quad (15\text{ dB})$$

$$h_{PA} = k \cdot 96\%$$

hHCA – Third Harmonic Peaking



Third Harmonic Peaking – Circuit

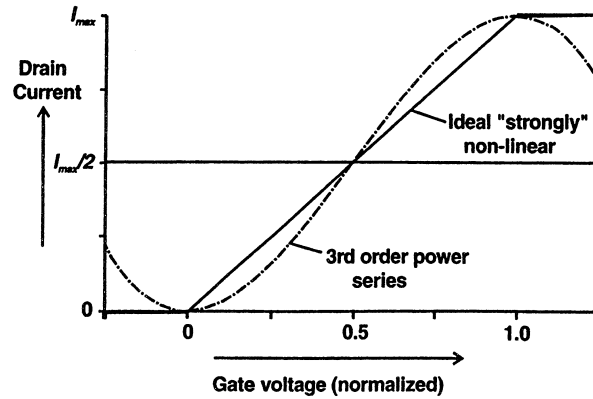


$$h_D = k \cdot 91\%$$

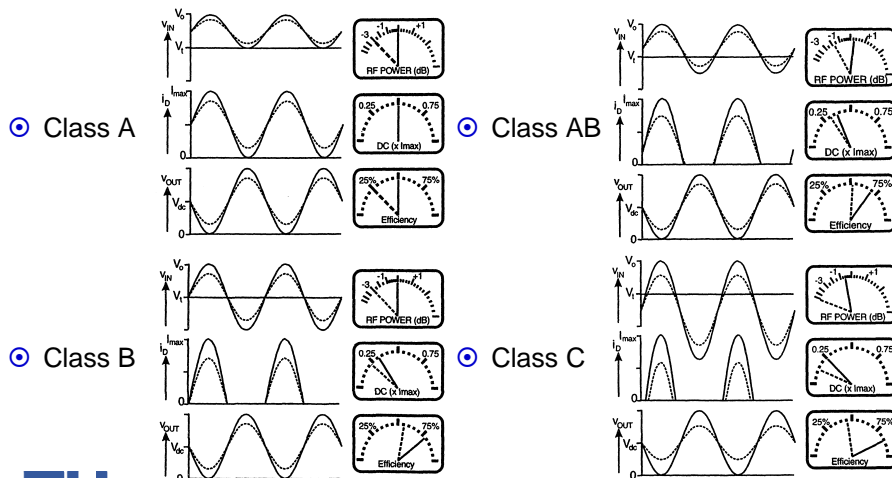
$$G = G_A + 0.6\text{dB} \quad (14.6\text{dB})$$

$$h_{PA} = k \cdot 87\%$$

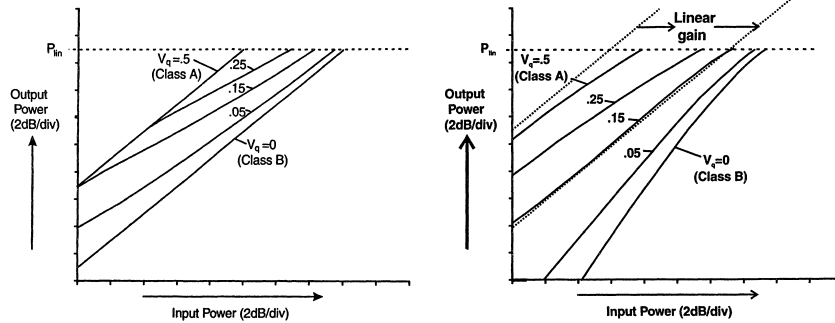
Linearity Aspects



Linearity Aspects



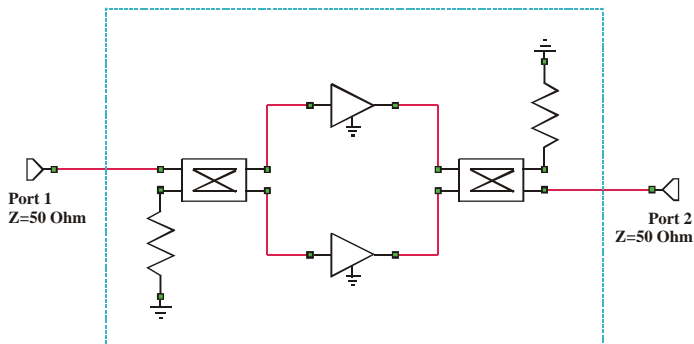
Linearity Aspects



- ⊙ Ideal strongly nonlinear model
- ⊙ Strong-weak nonlinear model

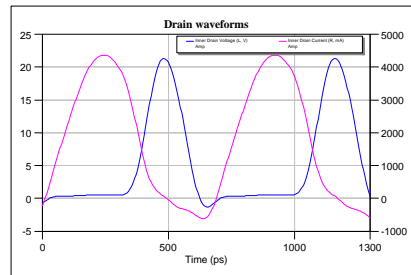
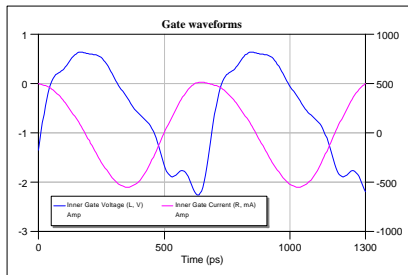
Amplifier Design – An Example

- ⊙ Balanced Amplifier Configuration



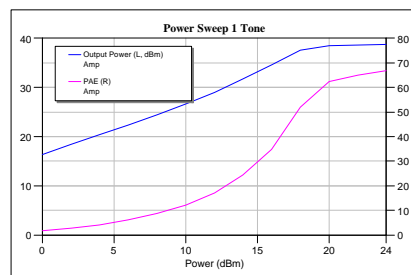
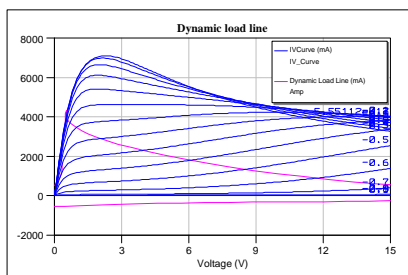
Amplifier Design – Simulation

Gate & Drain Waveforms



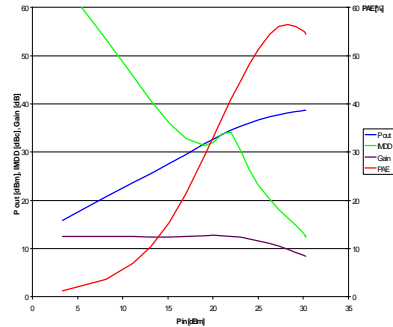
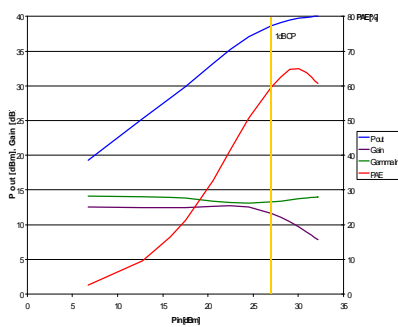
Amplifier Design – Simulation

Dynamic Load Line & Power Sweep



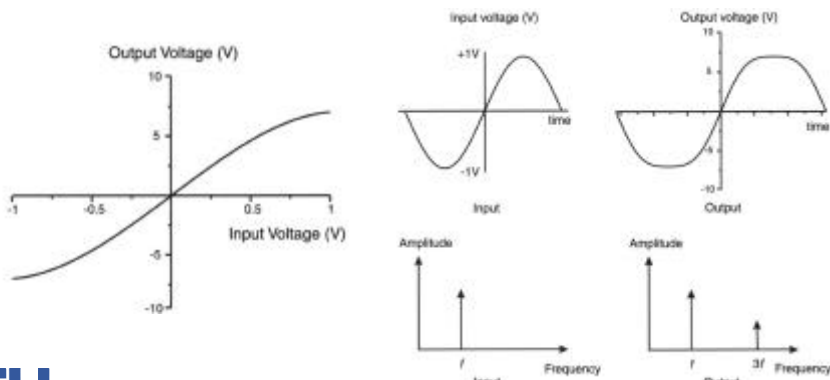
Amplifier Design – Measurements

Single Tone & Two Tone



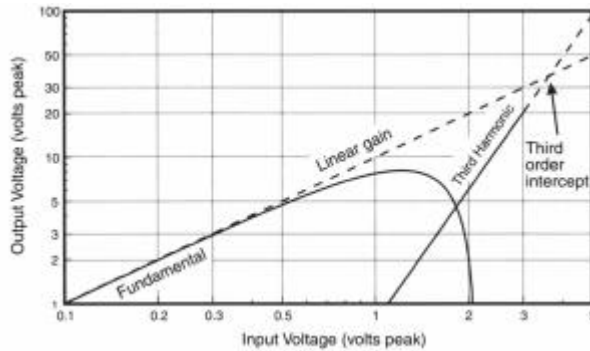
Amplifier Nonlinearity

- Gain and Phase depends on Input Signal
- 3rd Order Gain-Nonlinearities:



Amplifier Nonlinearity

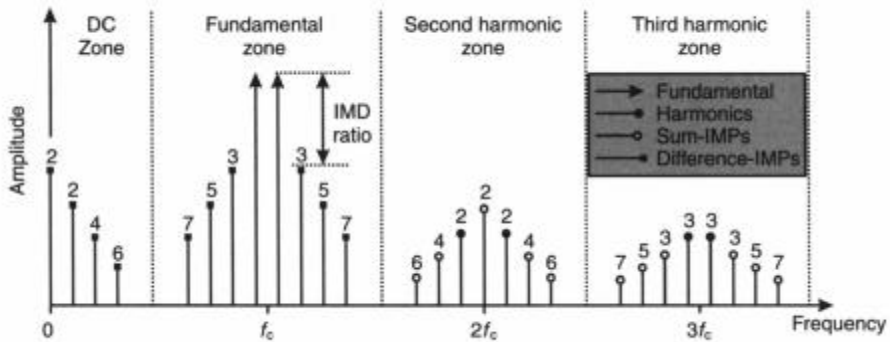
- Higher Output Level (close to Saturation) results in more Distortion/Nonlinearity



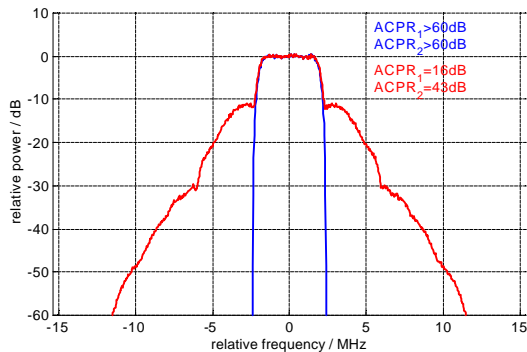
Nonlinearity leads to?

- Generation of Harmonics
- Intermodulation Distortion / Spectral Regrowth
- SNR (NPR) Degradation
- Constellation Deformation

Intermodulation and Harmonics



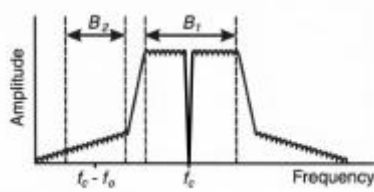
Spectral Regrowth



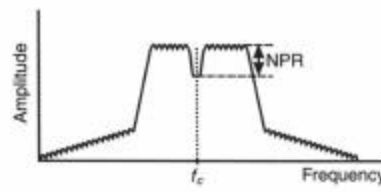
- ⊙ Energy in adjacent Channels
- ⊙ ACPR (Adjacent Channel Leakage Power Ratio) increases

Reduced NPR (Noise Power Ratio)

- Input Signal



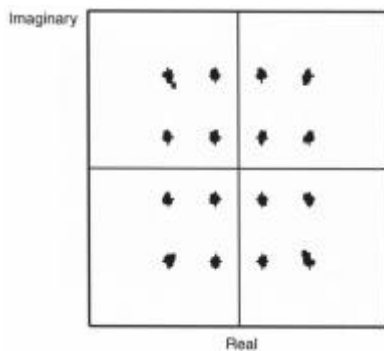
- Output Signal of Nonlinear Amplifier



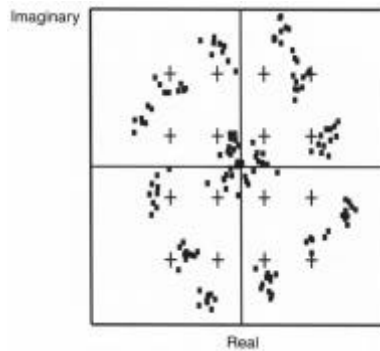
- Degradation of Inband SNR
- „Noisy“ Constellation

Constellation Deformation

- Input Signal



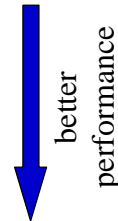
- Output Signal of Nonlinear Amplifier (with Gain- and Phase-Distortion)



Modeling of Nonlinearities

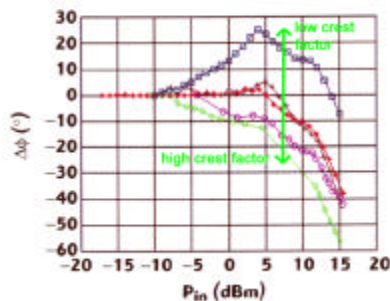
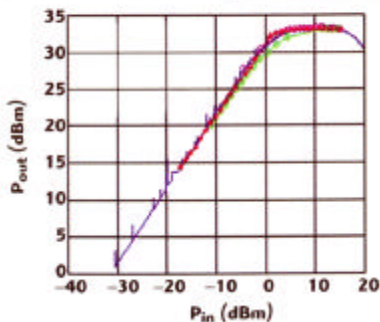
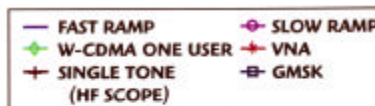
- with Memory-Effects
 - Volterra Series (= „Taylor Series with Memory“)

- without Memory-Effects
 - Saleh Model $f(r) = \frac{a_a r}{1 + b_a r^2}$ $g(r) = \frac{a_\theta r^2}{1 + b_\theta r^2}$
 - Taylor Series
 - Blum and Jeruchim Model
 - AM/AM- and AM/PM-conversion



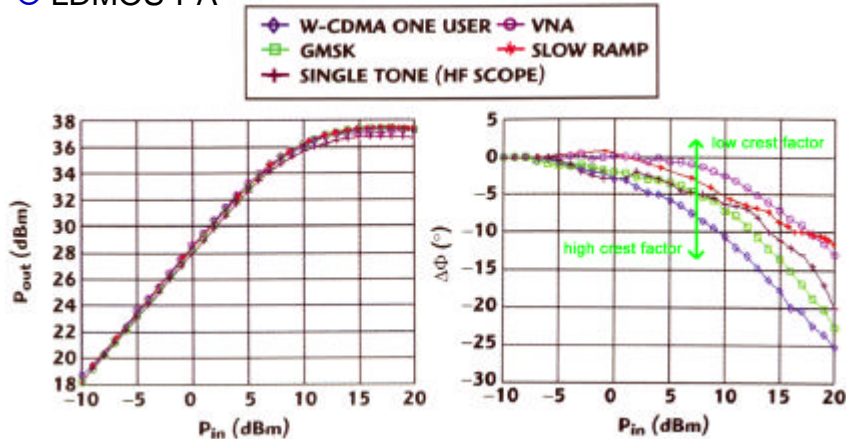
AM/AM- and AM/PM-Conversion

- GaAs-PA



AM/AM- and AM/PM-Conversion

◉ LD MOS-PA



How to preserve Linearity?

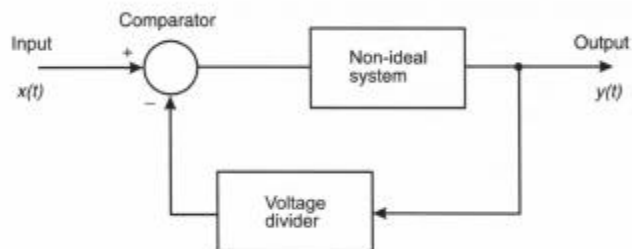
- ◉ Backed-Off Operation of PA
 - Simplest Way to achieve Linearity

- ◉ Linearity improving Concepts
 - Predistortion
 - Feedforward
 - ...

How to preserve Efficiency?

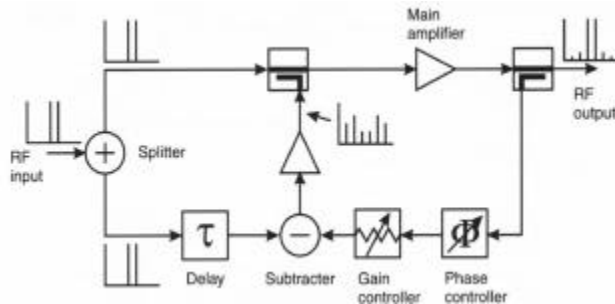
- ⊙ Efficiency improving Concepts
 - Doherty
 - Envelope Elimination and Restoration
 - ...
- ⊙ Linearity improving Concepts
 - Higher Linearity at constant Efficiency
→ Higher Efficiency at constant Linearity

Direct (RF) Feedback



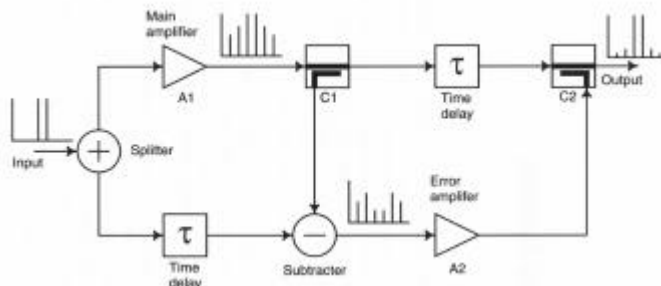
- ⊙ Classical Method
- ⊙ Decrease of Gain → Low Efficiency
- ⊙ Feedback needs more Bandwidth than Signal
- ⊙ Stability Problems at high Bandwidths

Distortion Feedback



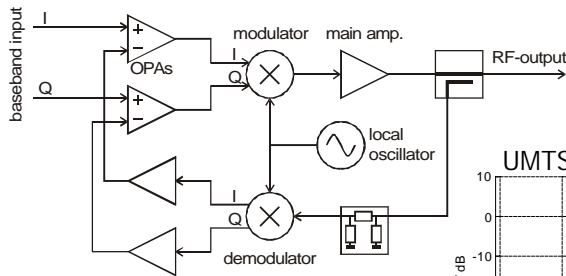
- ⦿ Feedback of outband Products only
- ⦿ Higher Gain than RF feedback
- ⦿ Stability Problems due to Reverse Loop

Feedforward

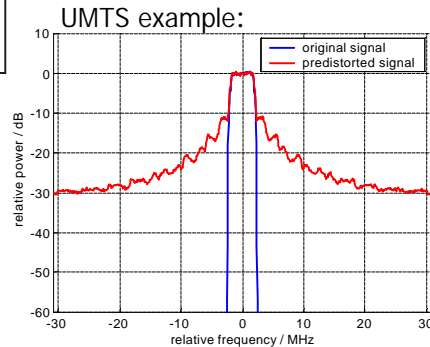


- ⦿ Overcomes Stability Problem by forward-only Loops
- ⦿ Critical to Gain/Phase-Imbalances
 - 0.5dB Gain Error \rightarrow -31dB Cancellation
 - 2.5° Phase Error \rightarrow -27dB Cancellation
- ⦿ Well suited for narrowband application

Cartesian Feedback

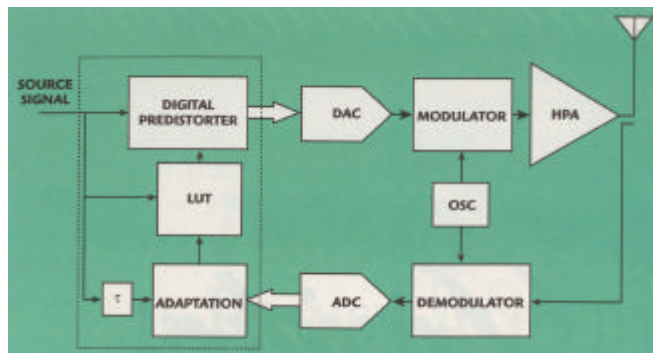


- ⊙ AM/AM- and AM/PM-correction
- ⊙ High Feedback-Bandwidth
- ⊙ Stability Problems

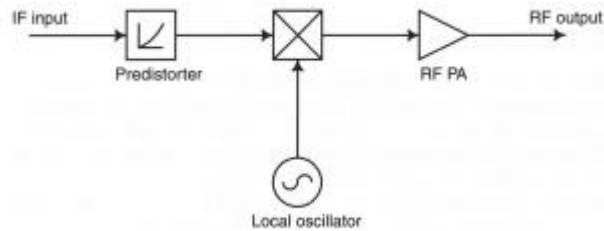


Digital Predistortion

- ⊙ Digital Implementation of „Cartesian Feedback“
- ⊙ Additional ADCs, DSP Power, Oversampling needed
- ⊙ Loop can be opened → no Stability Problems



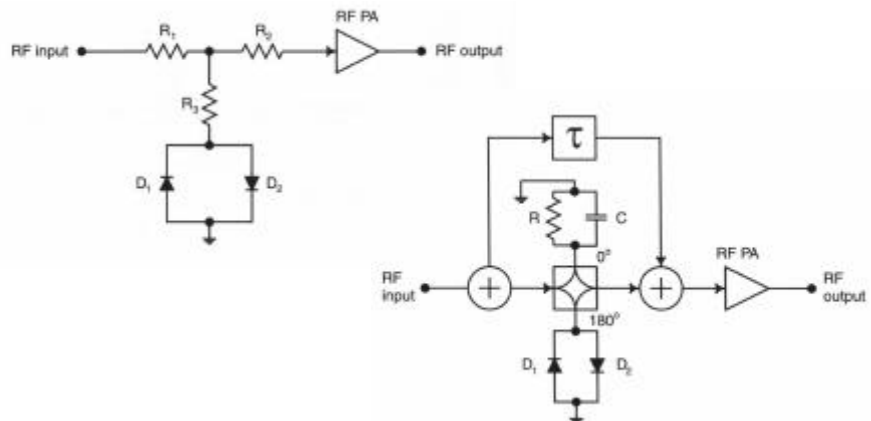
Analog Predistortion



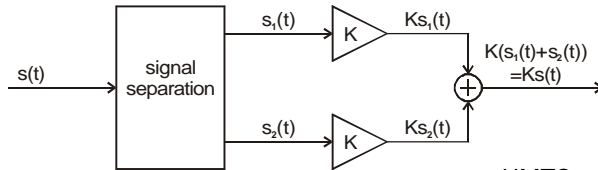
- ⊙ Predistorter has inverse Function of Amplifier
- ⊙ Leads to infinite Bandwidth (!)
- ⊙ Hard to realize (accuracy)

Analog Predistortion

- ⊙ Possible Realizations:

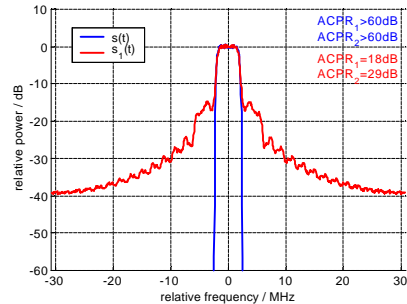


LINC (Linear Amplification by Nonlinear Components)



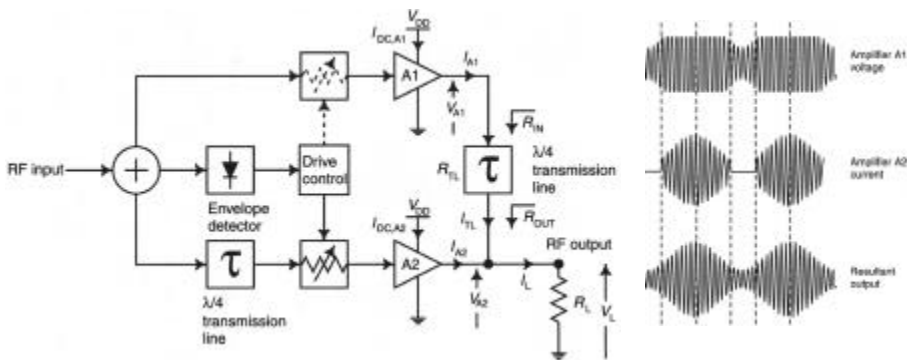
- ⊙ AM/AM- and AM/PM-correction
- ⊙ Digital separation required (accuracy!)
- ⊙ High Bandwidth, oversampling necessary
- ⊙ Stability guaranteed

UMTS example:



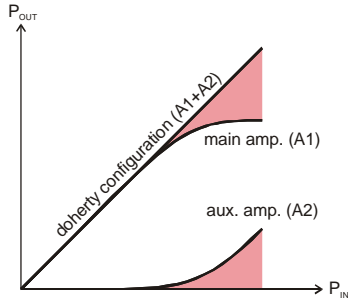
Doherty Amplifier

- ⊙ Auxiliary amplifier supports main amplifier during saturation
- ⊙ PAE can be kept high over a 6dB range

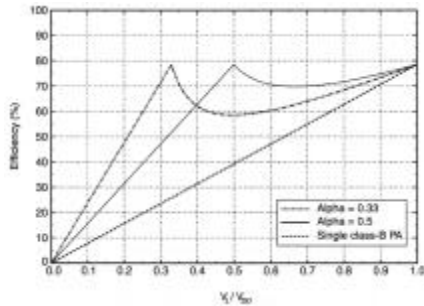


Doherty Amplifier

Gain vs. Input Power



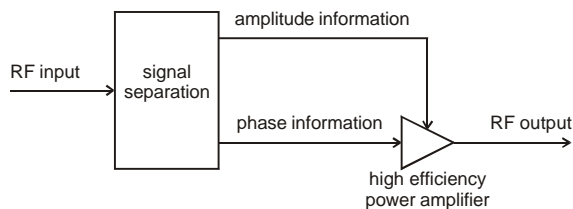
Efficiency vs. Input Power



- ⊙ No improvement of AM/AM- and AM/PM-distortion
- ⊙ Behavior of auxiliary amplifier very hard (impossible) to realize
- ⊙ Stability guaranteed

EER (Envelope Elimination and Restoration)

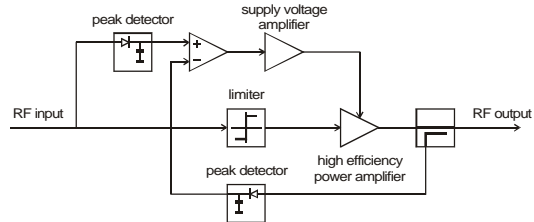
- ⊙ Separating phase and magnitude information
- ⊙ Elimination of AM/AM-distortion
- ⊙ Application of high-efficient amplifiers (independent of amplitude distortion)
- ⊙ Stability guaranteed



EER (Envelope Elimination and Restoration)

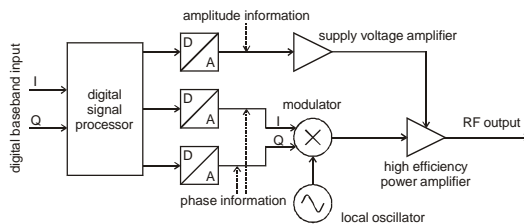
◉ Analog realization

- Limiter hard to build
- Accuracy problems
- Feedback necessary



◉ Digital realization

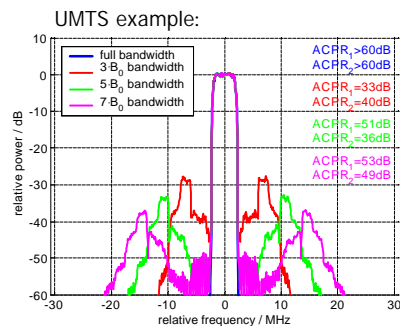
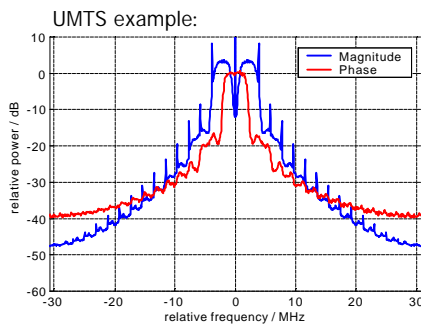
- Oversampling + high D/A-conversion rates required
- High power consumption of DSP and D/A-converters
- Possible feedback elimination
- Compensation of AM/PM-distortion possible



EER (Envelope Elimination and Restoration)

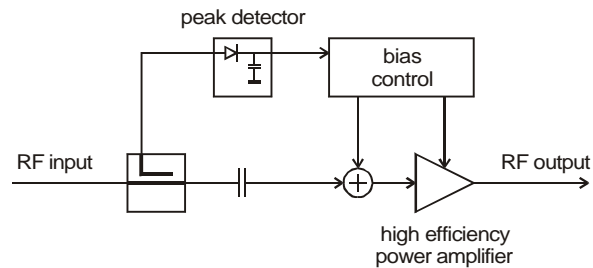
- ◉ Bandwidth of Magnitude- and phase-signal have higher than transmit signal

- ◉ Five times (!) oversampling necessary to achieve standard requirements



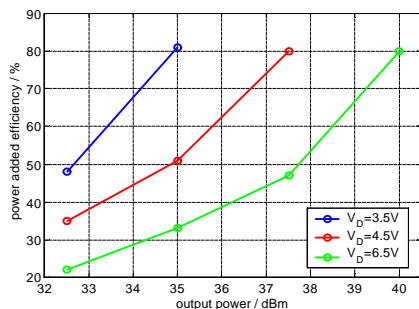
Adaptive Bias

- ⊙ Varying/Switching of Bias-Voltage depending on Input Power Level
- ⊙ Selection of Operating Point with high PAE
- ⊙ Applicably for nearly each type of Amplifier



Adaptive Bias

- ⊙ Single tone PAE for switched V_{DD} with V_G kept constant
- ⊙ Simply to implement Concept
- ⊙ Stability guaranteed
- ⊙ Possible problems:
 - DC-DC converter with high efficiency necessary
 - Possible Linearity Change (can increase and decrease) especially for HCAs



Summary


- ⊙ Digital Realization required to achieve Accuracy
- ⊙ Problem of Stability for high Bandwidth Application
- ⊙ Higher Bandwidths (Oversampling) necessary, depending on Order of IMD cancellation
- ⊙ Predistortion gives best Results while keeping Efficiency high (valid for high Output Levels > 40dBm)

Figure References

- ⊙ F. Zavosh et al,
“Digital Predistortion Techniques for RF Power Amplifiers with CDMA Applications”,
Microwave Journal, Oct. 1999
- ⊙ Peter B. Kenington,
“High-Linearity RF Amplifier Design”,
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- ⊙ Steve C. Cripps,
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